

Smart High-Side Power Switch

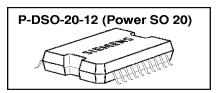
Two Channels: 2 x $30m\Omega$

Current Sense

Product Summary

| Operating Voltage | $V_{bb(on)}$ | 5.034V | | |
|---------------------|---------------------|--------------|--------------|--|
| | Active channels: | one | two parallel | |
| On-state Resistance | R _{oN} | $30 m\Omega$ | 15m Ω | |
| Load Current (ISO) | I _{L(ISO)} | 12 A | 24A | |
| Current Limitation | I _{L(SCr)} | 24A | 24A | |

Package



General Description

- N channel vertical power MOSFET with charge pump, ground referenced CMOS compatible input, diagnostic feedback and proportional load current sense monolithically integrated in Smart SIPMOS® technology.
- Fully protected by embedded protection functions

Applications

- μC compatible high-side power switch with diagnostic feedback for 12V and 24V grounded loads
- All types of resistive, inductive and capacitve loads
- Most suitable for loads with high inrush currents, so as lamps
- Replaces electromechanical relays, fuses and discrete circuits

Basic Functions

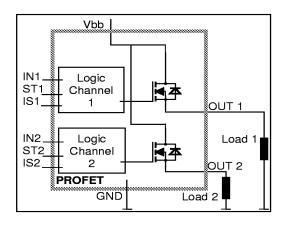
- CMOS compatible input
- Undervoltage and overvoltage shutdown with auto-restart and hysteresis
- Fast demagnetization of inductive loads
- Logic ground independent from load ground

Protection Functions

- Short circuit protection
- Overload protection
- Current limitation
- Thermal shutdown
- Overvoltage protection (including load dump) with external resistor
- Reverse battery protection with external resistor
- Loss of ground and loss of V_{bb} protection
- Electrostatic discharge protection (ESD)

Diagnostic Functions

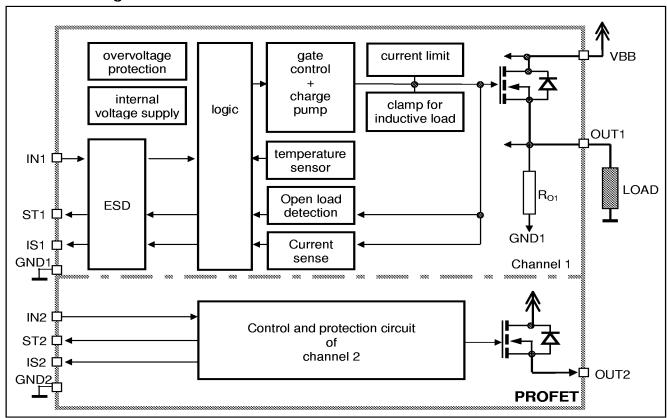
- Proportinal load current sense
- Diagnostic feedback with open drain output
- Open load detection in OFF-state with external resistor
- Feedback of thermal shutdown in ON-state



Data Sheet 1 1999-06-16



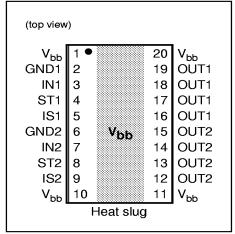
Functional diagram



Pin Definitions and Functions

| Pin | Symbol | Function |
|-----------------|-----------------|--|
| 1,10, 11,12, | V _{bb} | Positive power supply voltage. For high current applications the heat slug should be |
| | | used as Vbb connection. |
| 3 | IN1 | Input 1,2, activates channel 1,2 in case of |
| 7 | IN2 | logic high signal |
| 16,17, | OUT1 | Output 1,2, protected high-side power output |
| 18,19 | | of channel 1,2. All pins of each output have to |
| 12,13, | OUT2 | be connected in parallel for operation |
| 14,15 | | according ths spec (e.g. kilis). Design the |
| | | wiring for the max. short circuit current |
| 4 | ST1 | Diagnostic feedback 1,2 of channel 1,2 |
| 8 | ST2 | open drain, invers to input level |
| 2 | GND1 | Ground 1,2 of chip channel 1,2 |
| 6 | GND2 | |
| 5 | IS1 | Sense current output 1,2; proportional to the |
| 9 | IS2 | load current, zero in the case of current |
| | | limitation of the load current |
| Heatslug | V _{bb} | Positiv powersupply voltage. Good way to |
| | | design a very low thermal resistance. |

Pin configuration





Maximum Ratings at $T_j = 25$ °C unless otherwise specified

| Parameter | Symbol | Values | Unit |
|---|---|---------------------|------|
| Supply voltage (overvoltage protection see page 4) | $V_{ m bb}$ | 43 | V |
| Supply voltage for full short circuit protection $T_{j,\text{start}} = -40 \dots + 150^{\circ}\text{C}$ | $V_{ m bb}$ | 34 | V |
| Load current (Short-circuit current, see page 5) | <i>I</i> _ | self-limited | Α |
| Load dump protection ¹⁾ $V_{\text{LoadDump}} = V_{\text{A}} + V_{\text{S}}, \ V_{\text{A}} = 13.5 \text{ V}$ $R_{\text{I}^2} = 2 \Omega, \ t_{\text{d}} = 200 \text{ ms}; \ \text{IN} = \text{low or high,}$ each channel loaded with $R_{\text{L}} = 7.0 \Omega,$ | V _{Load dump³⁾} | 60 | V |
| Operating temperature range Storage temperature range | T _j T _{stg} | -40+150 -55+150 | °C |
| Power dissipation (DC) ⁴⁾ $T_a = 25^{\circ}\text{C}$: (all channels active) $T_a = 85^{\circ}\text{C}$: | P_{tot} | 3.8 2.0 | W |
| Maximal switchable inductance, single pulse $V_{bb} = 12V$, $T_{j,start} = 150^{\circ}C^{4}$, | | | |
| I_{L} = tbd A, E_{AS} = tbd mJ, 0Ω one channel: I_{L} = tbd A, E_{AS} = tbd mJ, 0Ω two parallel channels: see diagrams on page 10 | Z _L | tbd tbd | mH |
| Electrostatic discharge capability (ESD) IN: (Human Body Model) ST, IS: out to all other pins shorted: acc. MIL-STD883D, method 3015.7 and ESD assn. std. S5.1-1993 R=1.5k Ω ; C=100pF | V _{ESD} | 1.0 4.0 8.0 | kV |
| Input voltage (DC) | V_{IN} | -10 +16 | V |
| Current through input pin (DC) Current through status pin (DC) Current through current sense pin (DC) see internal circuit diagram page 9 | I _{IN} I _{ST} I _{IS} | ±2.0 ±5.0 ±14 | mA |

Thermal Characteristics

| Parameter and Conditions | | Symbol | Values | | | Unit |
|--------------------------|----------------------|-------------------|--------|-----|-----|------|
| | | | min | typ | max | |
| Thermal resistance | | | | | | |
| junction -case | each channel: | R _{thjs} | | | 1 | K/W |
| junction - ambient4) | one channel active: | $R_{\rm thia}$ | | 40 | | |
| - | all channels active: | _ | | 33 | | |

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Supply voltages higher than $V_{bb(AZ)}$ require an external current limit for the GND and status pins a 150 Ω resistor for the GND connection is recommended.

²⁾ R_1 = internal resistance of the load dump test pulse generator

³⁾ V_{Load dump} is set up without the DUT connected to the generator per ISO 7637-1 and DIN 40839

⁴⁾ Device on 50mm*50mm*1.5mm epoxy PCB FR4 with 6cm² (one layer, 70µm thick) copper area for V_{bb} connection. PCB is vertical without blown air.



Electrical Characteristics

| Parameter and Conditions, each of the two channels | Symbol | | Values | | Unit |
|--|---------------------------|------|--------|------------|-------|
| at $T_j = -40+150$ °C, $V_{bb} = 12$ V unless otherwise specified | | min | typ | max | |
| Load Switching Capabilities and Characteristics | ; | | | | |
| On-state resistance (V _{bb} to OUT); I _L = 5 A | | | | | |
| each channel, $T_i = 25^{\circ}C$: | RON | | 27 | 30 | mΩ |
| $T_{\rm i} = 150^{\circ}{\rm C}$: | | | 54 | 60 | |
| two parallel channels, $T_i = 25^{\circ}C$: | | | 14 | 15 | |
| Output voltage drop limitation at small load | | | | | |
| currents, see page 14 | $V_{\rm ON(NL)}$ | | 50 | | mV |
| $I_L = 0.5 \text{ A}$ $T_j = -40 + 150 ^{\circ}\text{C}$: | J 7 1 (1 1 2) | | | | |
| Nominal load current, ISO Norm | , | ايد | 40 | | |
| one channel active: | I _{L(NOM)} | 11 | 12 | | A |
| two parallel channels active: | | 22 | 24 | | |
| ISO 10483-1, 6.7: Von =0.5V T _C = 85°C | | | | | |
| Output current while GND disconnected or pulled up; Vbb = 30 V, VIN = 0, | / _{L(GNDhigh)} | | | 8 | mA |
| see diagram page 9; (not tested specified by design) | | | | | |
| Turn-on time ⁵⁾ IN \int to 90% V_{OUT} : | t _{on} | 25 | 70 | 150 | μs |
| Turn-off time IN \square to 10% V_{OUT} : | t _{off} | 25 | 80 | 200 | • |
| $R_{\rm L} = 12 \Omega$ | " | | | | |
| Slew rate on 5) | d V/dt _{on} | 0.1 | | 1 | V/µs |
| 10 to 30% V_{OUT} , $R_{\text{L}} = 12 \Omega$: | | | | | ' |
| Slew rate off ⁵⁾ | -d V/dt _{off} | 0.1 | | 1 | V/µs |
| 70 to 40% V_{OUT} , $R_{\text{L}} = 12 \Omega$: | | | | | |
| Operating Parameters | | | | | |
| Operating voltage ⁶⁾ | $V_{ m bb(on)}$ | 5.0 | [| 34 | V |
| Undervoltage shutdown | V _{bb(under)} | 3.2 | | 5.0 | V |
| Undervoltage restart $T_i = -40+25^{\circ}C$: | V _{bb(u rst)} | | 4.5 | 5.5 | v |
| $T_{\rm j}=+150^{\circ}{\rm C}$: | - bb(a ist) | | | 6.0 | |
| Undervoltage restart of charge pump | | | | | |
| see diagram page 13 \dot{T}_{i} =-40+25°C: T_{i} =150°C: | $V_{ m bb(ucp)}$ | | 4.7 | 6.5 7.0 | \ \ \ |
| Undervoltage hysteresis | 41/ | | 0.5 | | V |
| $\Delta V_{\text{obs}}(\text{under}) = V_{\text{obs}}(\text{urst}) - V_{\text{obs}}(\text{under})$ | $\Delta V_{ m bb(under)}$ | | 0.5 | | |
| Overvoltage shutdown | V _{bb(over)} | 34 | | 43 | V |
| Overvoltage restart | V _{bb(o rst)} | 33 | | | V |
| Overvoltage hysteresis | $\Delta V_{\rm bb(over)}$ | | 1 | | V |
| Overvoltage protection ⁷⁾ $T_i = -40$: | $V_{\rm bb(AZ)}$ | 41 | | | V |
| $I_{bb}=40 \text{ mA}$ $T_{bb}=+25+150 ^{\circ}\text{C}$: | | 43 | 47 | 52 | |

⁵⁾ See timing diagram on page 11.

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At supply voltage increase up to V_{bb} = 4.7 V typ without charge pump, $V_{OUT} \approx V_{bb}$ - 2 V



| Parameter and Conditions, each of the two channels | Symbol | Values | | | Unit |
|---|--------------------------|----------|------------|--------|------------|
| at T _j = -40+150°C, V_{bb} = 12 V unless otherwise specified | | min | typ | max | |
| | | | | | |
| Standby current ⁸⁾ $T_j = -40^{\circ} \text{C25}^{\circ} \text{C}$: | / _{bb(off)} | | 8 | 30 | μ A |
| $V_{IN} = 0$; see diagram page 10 $T_j = 150$ °C: | | | 24 | 50 | |
| Leakage output current (included in $I_{bb(off)}$) $V_{IN} = 0$ | $I_{L(off)}$ | | | 20 | μ A |
| Operating current 9, $V_{IN} = 5V$, | | | | | |
| $I_{\text{GND}} = I_{\text{GND1}} + I_{\text{GND2}},$ one channel on: two channels on: | / _{GND} | | 1.2 2.4 | 3 6 | mA |
| Protection Functions | | | | | |
| Current limit, (see timing diagrams, page 12) | | | | | |
| $T_{\rm j}$ =-40°C: | / _{L(lim)} | 48 | 56 | 65 | Α |
| $T_{\rm j}$ =25°C: | | 40 | 50 | 58 | |
| $T_{j} = +150^{\circ}C$: | | 31 | 37 | 45 | |
| Repetitive short circuit current limit, | | | | | |
| $T_{\rm j} = T_{\rm jt}$ each channel | /L(SCr) | | 24 | | Α |
| two parallel channels | | | 24 | | |
| (see timing diagrams, page 12) | | | | | |
| Initial short circuit shutdown time $T_{j,start} = 25^{\circ}C$: | toff(SC) | | 2.0 | | ms |
| (see timing diagrams on page 12) | | | | | |
| Output clamp (inductive load switch off) ¹⁰⁾ | 1 | ,, | | | V |
| at $V_{ON(CL)} = \dot{V}_{bb} - V_{OUT}$, $I_{L} = 40 \text{ mA}$ $T_{j} = -40^{\circ}\text{C}$: $T_{i} = 25^{\circ}\text{C}150^{\circ}\text{C}$: | $V_{ON(CL)}$ | 41 43 | 47 | 52 | V |
| Thermal overload trip temperature | T _{it} | 150 | | | °C |
| Thermal hysteresis | + · · | | 10 | | <u>- С</u> |
| Thermai hysteresis | ΔT_{jt} | | 10 | | |
| Reverse Battery | | | | | |
| Reverse battery voltage 11) | - $V_{ m bb}$ | | | 32 | V |
| Drain-source diode voltage ($V_{out} > V_{bb}$) $f_L = -4.0 \text{ A}, T_j = +150 ^{\circ} \text{C}$ | - <i>V</i> _{ON} | | 600 | | mV |

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Supply voltages higher than $V_{bb(AZ)}$ require an external current limit for the GND and status pins (a 150 Ω resistor in the GND connection is recommended). See also $V_{ON(CL)}$ in table of protection functions and circuit diagram page 9.

⁸⁾ Measured with load; for the whole device; all channels off

⁹⁾ Add I_{ST} , if $I_{ST} > 0$

 $^{^{10)}}$ If channels are connected in parallel, output clamp is usually accomplished by the channel with the lowest $^{
m V}_{
m ON(CL)}$

Requires a 150 Ω resistor in GND connection. The reverse load current through the intrinsic drain-source diode has to be limited by the connected load. Power dissipation is higher compared to normal operating conditions due to the voltage drop across the drain-source diode. The temperature protection is not active during reverse current operation! Input and Status currents have to be limited (see max. ratings page 3 and circuit page 9).



| Parameter and Conditions, each of the two channels | Symbol | | Values | ; | Unit |
|---|--------|-----|--------|-----|------|
| at T _j = -40+150°C, V_{bb} = 12 V unless otherwise specified | | min | typ | max | |

Diagnostic Characteristics

| Diagnostic Characteristics | | | | | |
|---|-----------------------|------|------|------|-----------|
| Current sense ratio ¹²⁾ , static on-condition, | | | | | |
| $V_{1S} = 05 \text{ V}, V_{bb(on)} = 6.5^{13})27\text{V},$ | | | | | |
| $k_{1}L_{1}S = l_{L} / l_{1}S$ $T_{j} = -40^{\circ}C, l_{L} = 5 A$: | k _{ILIS} | 4350 | 4800 | 5800 | |
| $T_{\rm j}$ = -40°C, $I_{\rm L}$ = 0.5 A: | | 3100 | 4800 | 7800 | |
| $T_{j}=25+150^{\circ}\text{C}, I_{L}=5 \text{ A}$: | | 4350 | 4800 | 5350 | |
| $T_{j}=25+150$ °C, $I_{L}=0.5$ A: | | 3800 | 4800 | 6300 | |
| Current sense output voltage limitation | | | | | |
| $T_{j} = -40 \dots + 150^{\circ}C$ /IS = 0, $I_{L} = 5 A$: | $V_{IS(lim)}$ | 5.4 | 6.1 | 6.9 | V |
| Current sense leakage/offset current | | | | | |
| $T_{j} = -40 \dots +150^{\circ}C$ $V_{N}=0, V_{S}=0, I_{L}=0$: | I _{IS(LL)} | 0 | | 1 | μA |
| $V_{IN}=5 \text{ V}, \ V_{IS}=0, \ I_{L}=0$: | I _{IS(LH)} | 0 | | 15 | • |
| VIN=5 V, $VIS=0$, $VOUT=0$ (short circuit) | I _{IS(SH)} | 0 | | 10 | |
| (/IS(SH) not tested, specified by design) | | | | | |
| Current sense settling time to I _{IS static} ±10% after | | | | | |
| positive input slope, $I_L = 0$ 5 A | $t_{\text{son(IS)}}$ | | | 300 | μs |
| (not tested, specified by design) | | | | | |
| Current sense settling time to 10% of I_{IS} static after | | | | | |
| negative input slope, $I_L = 5$ 0 A | t _{soff(IS)} | | 30 | 100 | μs |
| (not tested, specified by design) | | | | | |
| Current sense rise time (60% to 90%) after change of load current $I_L = 2.5$ 5 A | , | | 10 | | |
| (not tested, specified by design) | $t_{\rm slc(IS)}$ | | 10 | | μs |
| Open load detection voltage ¹⁴ (off-condition) | V _{OUT(OL)} | 2 | 3 | 4 | |
| Internal output pull down | 03.(OL) | _ | | - | |
| (pin 17,18 to 2 resp. 13,14 to 6), VOUT=5 V | R_{\circ} | 5 | 15 | 40 | $k\Omega$ |
| | | | | | |

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This range for the current sense ratio refers to all devices. The accuracy of the k_{ILIS} can be raised at least by a factor of two by matching the value of k_{ILIS} for every single device.

In the case of current limitation the sense current $l_{\rm IS}$ is zero and the diagnostic feedback potential $V_{\rm ST}$ is High. See figure 2c, page 12.

¹³⁾ Valid if $V_{
m bb(u\ rst)}$ was exceeded before.

¹⁴⁾ External pull up resistor required for open load detection in off state.



| Parameter and Conditions, each of the two channels | ameter and Conditions, each of the two channels Symbol Values | | Values | | Unit |
|---|---|-----|--------|-----|------|
| at $T_j = -40+150$ °C, $V_{bb} = 12$ V unless otherwise specified | | min | typ | max | |

Input and Status Feedback¹⁵⁾

| Input resistance (see circuit page 9) | R_{I} | 3.0 | 4.5 | 7.0 | kΩ |
|---|------------------------|-----|-----|------------|----|
| Input turn-on threshold voltage | $V_{IN(T+)}$ | | | 3.5 | V |
| Input turn-off threshold voltage | $V_{IN(T-)}$ | 1.5 | | | V |
| Input threshold hysteresis | $\Delta V_{\rm IN(T)}$ | | 0.5 | | V |
| Off state input current $V_{IN} = 0.4 \text{ V}$: | I _{IN(off)} | 1 | | 50 | μΑ |
| On state input current $V_{IN} = 5 \text{ V}$: | I _{IN(on)} | 20 | 50 | 90 | μΑ |
| Delay time for status with open load after Input neg. slope (see diagram page 13) | t _{d(ST OL3)} | | 400 | | μs |
| Status delay after positive input slope (not tested, specified by design) | t _{don(ST)} | | 13 | | μs |
| Status delay after negative input slope (not tested, specified by design) | t _{doff(ST)} | | 1 | | μs |
| Status output (open drain) | | | | | |
| Zener limit voltage $T_j = -40 + 150$ °C, $I_{ST} = +1.6$ mA: | $V_{ m ST(high)}$ | 5.4 | 6.1 | 6.9 | V |
| ST low voltage $T_i = -40+25$ °C, $I_{ST} = +1.6$ mA: $T_j = +150$ °C, $I_{ST} = +1.6$ mA: | $V_{\rm ST(low)}$ | | | 0.4 0.7 | |
| Status leakage current, $V_{ST} = 5 \text{ V}$, $T_j=25 \dots +150 ^{\circ}\text{C}$: | I _{ST(high)} | | | 2 | μΑ |

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 $^{^{\}rm 15)}\,$ If ground resistors ${\rm R}_{\rm GND}$ are used, add the voltage drop across these resistors.



Truth Table

| | Input 1 | Output 1 | Status 1 | Current Sense 1 |
|-------------------------------|---------|-------------------|-----------------------|-----------------------------------|
| | Input 2 | Output 2 | Status 2 | Current Sense 2 |
| | level | level | level | l _{IS} |
| Normal | L | L | Н | 0 |
| operation | Н | Н | L | nominal |
| Current- | L | L | Н | 0 |
| limitation | H | Н | Н | 0 |
| Short circuit to | L | L | Н | 0 |
| GND | Н | L ¹⁶) | Н | 0 |
| Over- | L | L | Н | 0 |
| temperature | Н | L | Н | 0 |
| Short circuit to | L | Н | L ¹⁷) | 0 |
| V _{bb} | Н | Н | L | <nominal 18)<="" td=""></nominal> |
| Open load | L | L ¹⁹) | H (L ²⁰⁾) | 0 |
| | н | Н | L ´ | 0 |
| Undervoltage | L | L | Н | 0 |
| | н | L | L | 0 |
| Overvoltage | L | L | Н | 0 |
| | н | L | L | 0 |
| Negative output voltage clamp | L | L | Н | 0 |

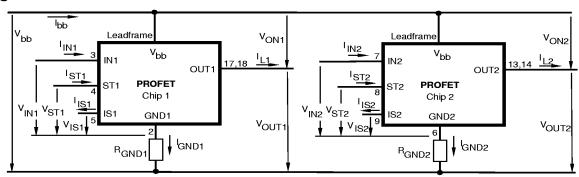
L = "Low" Level

X = don't care

Z = high impedance, potential depends on external circuit

H = "High" Level Status signal after the time delay shown in the diagrams (see fig 5. page 13) Parallel switching of channel 1 and 2 is possible by connecting the inputs and outputs in parallel. The status outputs ST1 and ST2 have to be configured as a 'Wired OR' function with a single pull-up resistor. The current sense outputs IS1 and IS2 have to be connected with a single pull-down resistor.

Terms



Leadframe (V_{bb}) is connected to pin 1,10,11,12,15,16,19,20

External R_{GND} optional; two resistors R_{GND1} , R_{GND2} = 150 Ω or a single resistor R_{GND} = 75 Ω for reverse battery protection up to the max. operating voltage.

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The voltage drop over the power transistor is $V_{\rm bb}$ - $V_{\rm OUT}$ > 3V typ. Under this condition the sense current $I_{\rm IS}$ is zero

An external short of output to V_{bb} , in the off state, causes an internal current from output to ground. If R_{GND} is used, an offset voltage at the GND and ST pins will occur and the $V_{ST low}$ signal may be errorious.

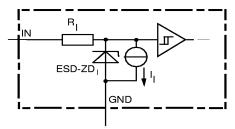
¹⁸⁾ Low ohmic short to $V_{
m bb}$ may reduce the output current $I_{
m L}$ and therefore also the sense current $I_{
m IS}$.

¹⁹⁾ Power Transistor off, high impedance

 $^{^{20)}}$ with external resistor between $V_{\tiny BB}$ and OUT

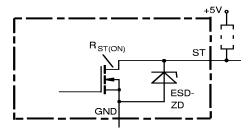


Input circuit (ESD protection), IN1 or IN2



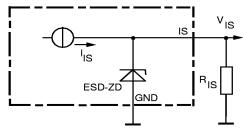
The use of ESD zener diodes as voltage clamp at DC conditions is not recommended.

Status output, ST1 or ST2



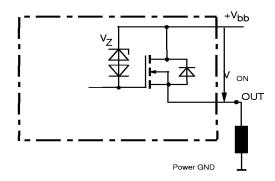
ESD-Zener diode: 6.1 V typ., max 5.0 mA; $R_{ST(ON)}$ < 375 Ω at 1.6 mA. The use of ESD zener diodes as voltage clamp at DC conditions is not recommended.

Current sense output



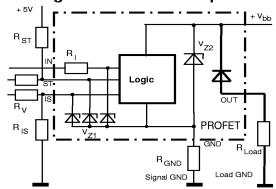
ESD-Zener diode: 6.1 V typ., max 14 mA; $R_{IS} = 1 \text{ k}\Omega$ nominal

Inductive and overvoltage output clamp, OUT1 or OUT2



 V_{ON} clamped to $V_{ON(CL)} = 47 \text{ V typ.}$

Overvoltage and reverse batt. protection

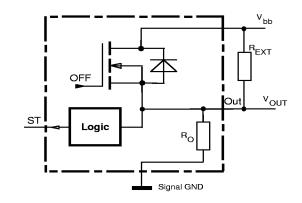


 $V_{Z1}=6.1~V$ typ., $V_{Z2}=47~V$ typ., $R_{GND}=150~\Omega$, $R_{ST}=15k\Omega$, $R_{I}=4.5k\Omega$ typ., $R_{IS}=1k\Omega$, $R_{V}=15k\Omega$, In case of reverse battery the current has to be limited by the load. Temperature protection is not active

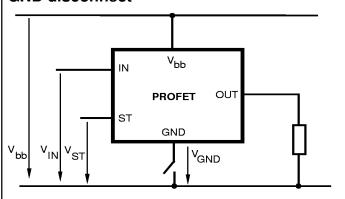
Open-load detection OUT1 or OUT2

OFF-state diagnostic condition:

 $V_{\text{OUT}} > 3 \text{ V typ.}$; IN low



GND disconnect

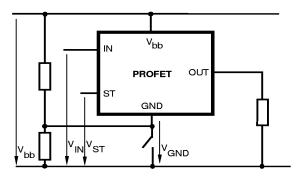


Any kind of load. In case of IN = high is $V_{OUT} \approx V_{IN} - V_{IN(T+)}$. Due to $V_{GND} > 0$, no $V_{ST} =$ low signal available.

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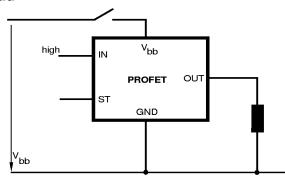


GND disconnect with GND pull up



Any kind of load. If $V_{GND} > V_{IN} - V_{IN(T+)}$ device stays off Due to $V_{GND} > 0$, no $V_{ST} = low$ signal available.

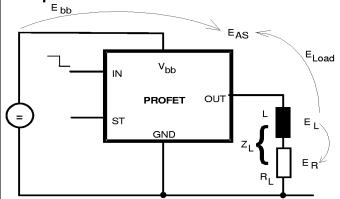
V_{bb} disconnect with energized inductive load



For inductive load currents up to the limits defined by Z_L (max. ratings and diagram on page 10) each switch is protected against loss of V_{bb} .

Consider at your PCB layout that in the case of Vbb disconnection with energized inductive load all the load current flows through the GND connection.

Inductive load switch-off energy dissipation



Energy stored in load inductance:

$$E_{L} = \frac{1}{2} \cdot L \cdot I_{L}^{2}$$

While demagnetizing load inductance, the energy dissipated in PROFET is

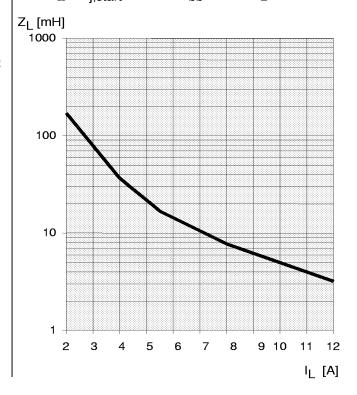
$$E_{AS} = E_{bb} + E_L - E_R = \int V_{ON(CL)} \cdot i_L(t) dt$$

with an approximate solution for $R_L > 0 \Omega$:

$$E_{AS} = \frac{I_L \cdot L}{2 \cdot R_L} \left(V_{bb} + |V_{OUT(CL)}| \right) ln \left(1 + \frac{I_L \cdot R_L}{|V_{OUT(CL)}|} \right)$$

Maximum allowable load inductance for a single switch off (one channel)⁴⁾

$$L = f(I_L)$$
; T_{i.start} = 150°C, V_{bb} = 12 V, R_L = 0 Ω



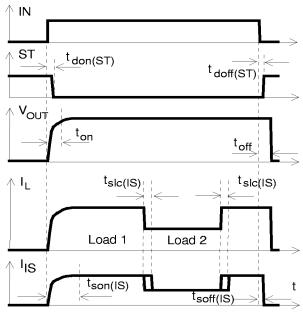
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Timing diagrams

Both channels are symmetric and consequently the diagrams are valid for channel 1 and channel 2

Figure 1a: Switching a resistive load, change of load current in on-condition:



The sense signal is not valid during settling time after turn or change of load current.

Figure 1b: V_{bb} turn on:

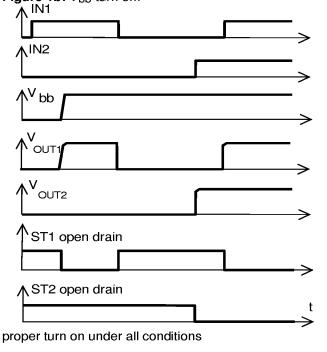


Figure 2a: Switching a resistive load, turn-on/off time and slew rate definition:

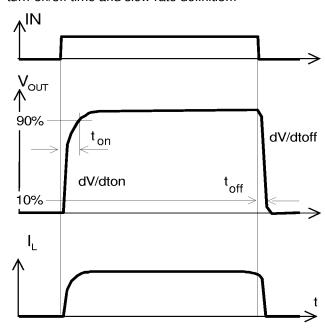
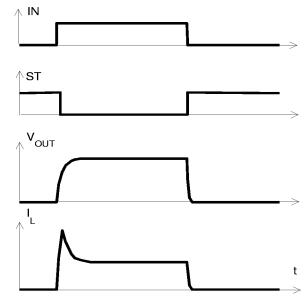


Figure 2b: Switching a lamp:



The initial peak current should be limited by the lamp and not by the current limit of the device.

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Figure 2c: Switching a lamp with current limit:

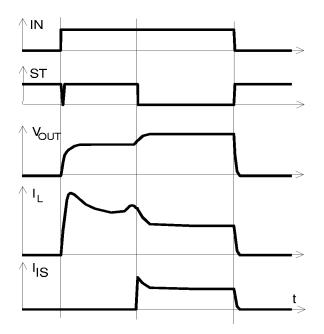
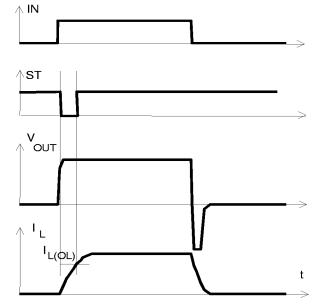
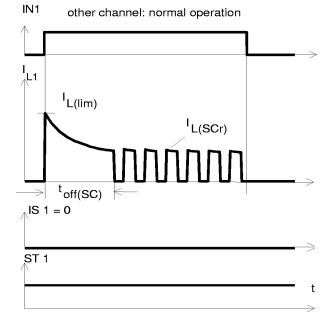


Figure 2d: Switching an inductive load



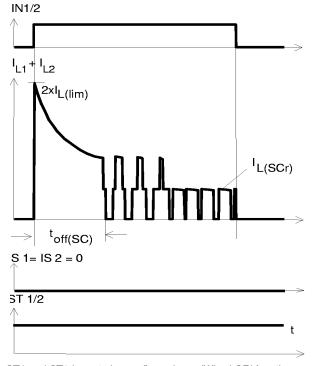
*) if the time constant of load is too large, open-load-status may occur

Figure 3a: Turn on into short circuit: shut down by overtemperature, restart by cooling



Heating up of the chip may require several milliseconds, depending on external conditions

Figure 3b: Turn on into short circuit: shut down by overtemperature, restart by cooling (two parallel switched channels 1 and 2)



ST1 and ST2 have to be configured as a 'Wired OR' function ST1/2 with a single pull-up resistor.



Figure 4a: Overtemperature: Reset if $T_i < T_{it}$

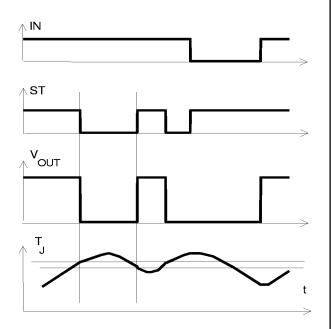


Figure 5a: Open load: detection (with R_{EXT}), turn on/off to open load

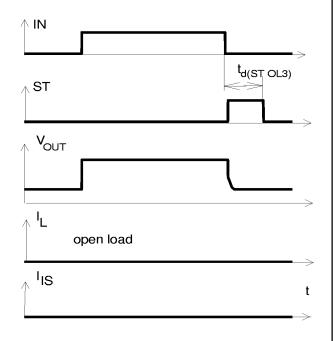


Figure 6a: Undervoltage:

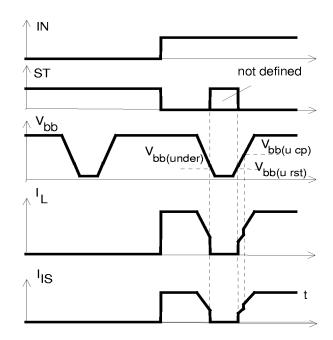
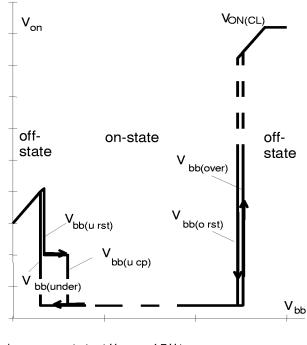


Figure 6b: Undervoltage restart of charge pump



charge pump starts at $V_{\rm bb(ucp)}$ =4.7 V typ.

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Figure 7a: Overvoltage:

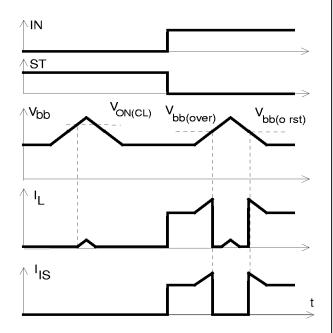
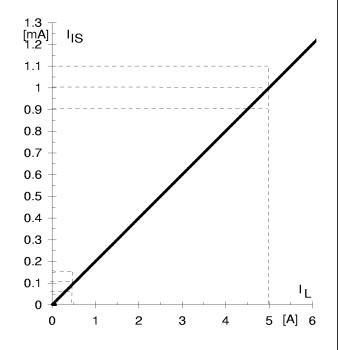


Figure 8a: Current sense versus load current²¹::



This range for the current sense ratio refers to all devices. The accuracy of the k_{ILIS} can be raised at least by a factor of two by matching the value of k_{ILIS} for every single device.

Figure 8b: Current sense ratio:

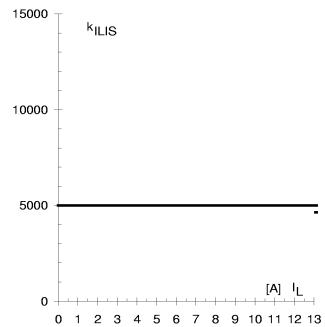
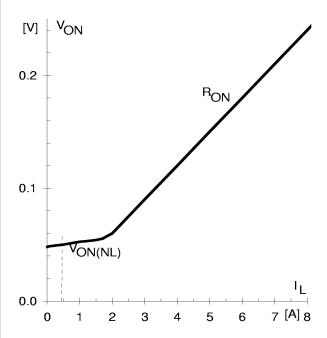


Figure 9a: Output voltage drop versus load current:



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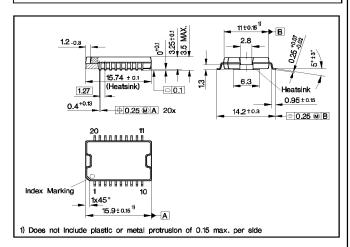


Package and Ordering Code

Standard: P-DSO-20-12 (Power SO 20)

| Sales Code | BTS 840 |
|---------------|---------|
| Ordering Code | tbd |

All dimensions in millimetres



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